Introduction

We consider the detection of visual evoked potentials (VEP's) with the aim of developing a system for objective measurement of visual acuity. A matched subspace filter (MSF) is demonstrated to outperform a number of other evoked potential detectors. The MSF is suitable for detecting multi-harmonic VEP's, unlike the earlier single-Fourier component detectors. The MSF has also been shown to be a uniformly most powerful detector for unknown signals in a given subspace with unknown noise.





Snellen Chart



20/200 acuity implies: visual system resolves 200/20=10 minutes of arc





Why objective acuity measurement?

- Snellen acuity measurements unreliable through first 3 years of life
- amblyopia can be reversed if treated early
- some patients have disturbed consciousness: mentally retarded, cerebral paulsey, head injury, alzheimer's





Objective Measurement of Visual Acuity







Overview of Detection Algorithms

Notation:

$$x = \begin{bmatrix} x_1^T & x_2^T & \cdots & x_L^T \end{bmatrix}^T \sim M \times 1$$

 x_i : response to ith contrast reversal

Ensemble Average:

$$\overline{x} = \frac{1}{L} \sum_{k=1}^{L} x_k$$





Generalized T^2 Statistic (Picton et al., 1987)

DFT of ith response:

$$X_{i} = \begin{bmatrix} 1 & \cos(2\pi f_{k}) & \cdots & \cos(2\pi (N-1)f_{k}) \\ 1 & \sin(2\pi f_{k}) & \cdots & \sin(2\pi (N-1)f_{k}) \end{bmatrix} x_{i}$$

sample mean:

sample covariance:

$$\overline{X} = \frac{1}{L} \sum_{k=1}^{L} X_k \qquad C = \frac{1}{L} \sum_{k=1}^{L} (X_k - \overline{X}) (X_k - \overline{X})^T$$





Generalized T^2 Statistic (cont.)

if X_i is $N(\mu, \Sigma)$

likelihood ratio test for:

$$H_0: \mu = 0 \quad \text{vs.} \quad H_1: \mu > 0$$
$$T^2 = L\overline{X}^T C^{-1} \overline{X}$$
$$\frac{L-2}{2(L-1)} T^2 \quad \text{has density:} \quad F(2, L-2)$$





Circular *T*² Statistic (Victor and Mast, 1991)

DFT:

$$Y_{i} = \begin{bmatrix} 1 & \exp(-j2\pi f_{k}) & \cdots & \exp(-j2\pi(N-1)f_{k}) \end{bmatrix} x_{i}$$

~ $N(\mu, \Sigma)$
 $H_{0}: \mu = 0 \text{ vs. } H_{1}: \mu > 0$
 $T_{circ}^{2} = (L-1)\frac{(L-1)|\overline{Y}|^{2}}{\sum_{k=1}^{L}|Y_{k} - \overline{Y}|^{2}}$ $\overline{Y} = \frac{1}{L}\sum_{k=1}^{L}Y_{k}$

 LT_{circ}^2 has density: F(2, 2L-2)





Rayleigh Phase Criterion (Kuwada et al, 1986)

$$R_{p} = \frac{1}{L} \sqrt{\left(\sum_{k=1}^{L} \cos \theta_{k}\right)^{2} + \left(\sum_{k=1}^{L} \sin \theta_{k}\right)^{2}}$$

compared against a table of thresholds for a given false alarm rate.





Nonparametric

- Looks at power in ensemble averages obtained using all possible sign permutations of each single trial.
- If the power in the average corresponding to all "+" signs is in the top 5% of all ensemble average powers, a detection is made.





Matched Subspace Filtering (Scharf, 1991)

Assmptions: • Signal *s* lies in a known subspace.

$$s = \sum_{k=1}^{2N} \alpha_k s_k$$

- Exact form of signal (α_k) is unknown.
- Additive white Gaussian noise power is unknown.

Want a statistic which is uniformly most powerful under these three constraints.





Matched Subspace Filtering (cont.)

MSF Statistic:

$$f = \frac{M - 2N}{2N} \frac{x^T P_s x}{x^T (I - P_s) x}$$
$$P_s = S \left(S^T S \right)^{-1} S^T$$

f has density
$$F(2N, M-2N)$$





VEP Signal Model

The VEP consists of *N* even harmonics of the contrast reversal frequency:

$$s_k(n) = \cos(2\pi f_k n), \ k = 1, 2, \dots, N$$
$$s_k(n) = \sin(2\pi f_{k-N} n), \ k = N+1, N+2, \dots, 2N$$
$$f_k = 2k \times f_{stim}, k = 1, 2, \dots, N$$

EEG noise is assumed to be an
$$AR(p)$$
 process.
-must prewhiten





Whitening Filter Design

AR(p) model:

$$z(n) = -\sum_{k=1}^{p} a_k z(n-k) + u(n)$$

u(n): Gaussian white noise

optimal whitening filter:

$$w = \begin{bmatrix} 1 & a_1 & a_2 & \cdots & a_p \end{bmatrix}^T$$





Yule-Walker Equations

$$R_{zz} \widetilde{w} = b$$

$$R_{zz} \left[\begin{matrix} r_{zz}(0) & r_{zz}(-1) & \cdots & r_{zz}(-p) \\ r_{zz}(1) & r_{zz}(0) & \cdots & r_{zz}(-(p-1)) \\ \vdots & \vdots & \ddots & \vdots \\ r_{zz}(p) & r_{zz}(p-1) & \cdots & r_{zz}(0) \end{matrix} \right]$$

$$\widetilde{w} = \beta \begin{bmatrix} 1 & a_1 & \cdots & a_p \end{bmatrix}^T \qquad b = \begin{bmatrix} 1 & 0 & \cdots & 0 \end{bmatrix}^T$$

$$\widetilde{w} = \beta w$$
 $\beta = \text{constant}$





Perturbation due to Single Sinusoid

$$(R_{zz} + R_{ss})(\widetilde{w} + \delta \widetilde{w}) = b$$

$$R_{ss} = \begin{bmatrix} 1 & \cos(2\pi f_0) & \cdots & \cos(2\pi f_0 p) \\ \cos(2\pi f_0) & 1 & \cdots & \cos(2\pi f_0 (p-1)) \\ \vdots & \vdots & \ddots & \vdots \\ \cos(2\pi f_0 p) & \cos(2\pi f_0 (p-1)) & \cdots & 1 \end{bmatrix}$$

bias: $\delta \widetilde{w} = \beta \delta w$





Perturbation Analysis

If: $||R_{zz}^{-1}|||R_{ss}|| < 1$

then (Stewart, 1973) $\frac{\|\mathbf{C}\|}{\|}$

$$\frac{\delta \widetilde{w}}{\left|\widetilde{w}\right|} \leq \frac{\left\|R_{zz}^{-1}\right\|}{1 - \left\|R_{zz}^{-1}\right\|} \left\|R_{ss}\right\|}$$

relative bias:

$$\frac{\left\|\delta w\right\|}{\left\|w\right\|} \leq \left\|R_{zz}^{-1}\right\| \left\|R_{ss}\right\|$$



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and if: $\|R_{zz}^{-1}\|\|R_{ss}\| \ll 1$



Relative Bias Bound vs. SNR







Estimated Relative Bias vs. SNR







Estimated PSD Before/After Prewhitening







Methods

- Vertical square wave gratings at 92% contrast.
- 5.5 degree circular field, luminance 30 foot lamberts.
- counterphase contrast reversal at 3.75 Hz, 7.5 reversals/s.
- spatial frequency varied from 4 c/d to 40 c/d, randomly.
- nineteen, 173-second runs at a fixed spatial frequency were obtained.
- EEG measured from O_z - C_z .





Data Analysis

- Each fixed spatial frequency run broken up into 25 M = 864-sample measurement vectors.
- Each measurement vector was filtered with a p = 15 whitening filter.
- Probability of detection (PD) was estimated for each spatial frequency for:
 - RPC
 T²
 T²_{circ}
 ROTP
 MSF (N = 4)





Mean PD vs. Spatial Frequency







Conclusions

- Objective acuity measurement requires accurate and sensitive VEP detection.
- MSF detector looks at several harmonics of contrast reversal frequency, has better performance than previous detectors.
- Prewhitening does not affect signal component at low SNR.



